



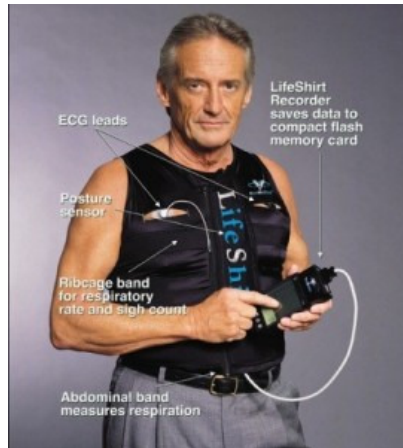
FACULTY OF ENGINEERING AND
ARCHITECTURE

Novel antenna design paradigms for the Internet of Things

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- **The Internet-of-Things**
 - **opportunities and design challenges**
- **Stochastic antenna design framework**
 - Production uncertainties
 - Substrate compression
 - Substrate bending
- **Some novel IoT antenna designs**
 - Substrate-Integrated-Waveguide (SIW)
Cavity-Backed Slot (CBS) topology
 - Three-element antenna array for integration into furniture
 - Half-mode SIW CBS antenna on cork substrate
- **Conclusions**

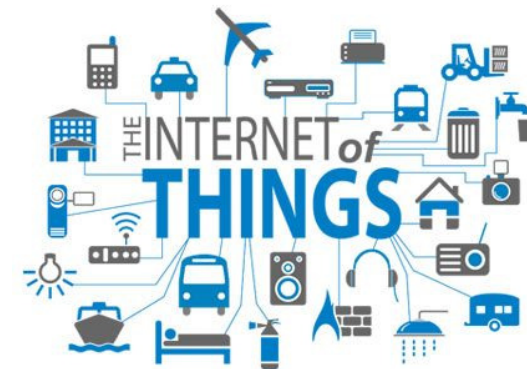
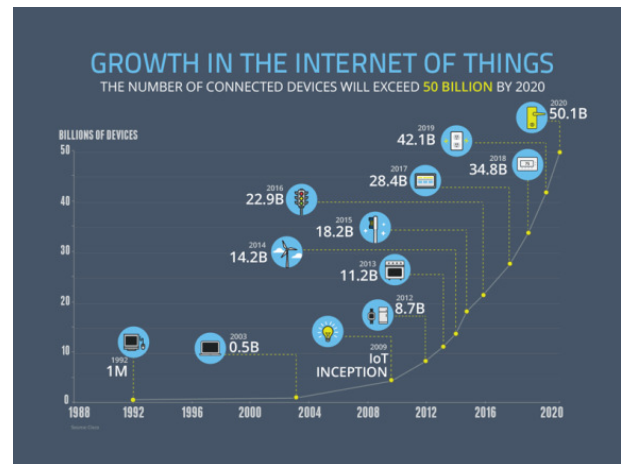
Invisibly integrated wireless communication systems

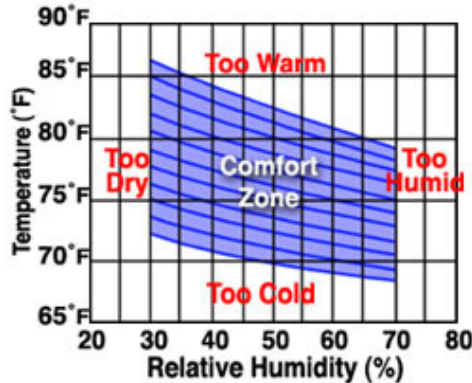


Increased Functionality



High data rates





Stable antenna performance requires taking into account adverse conditions during design phase:

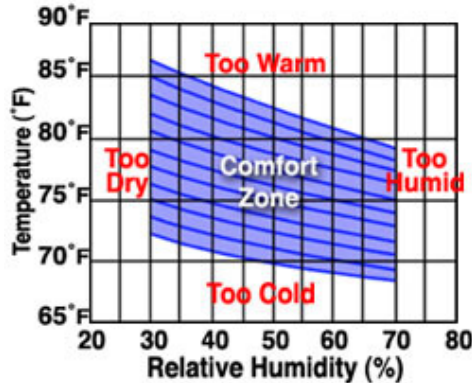
- effect of varying environmental conditions
- effect of fabrication tolerances
- effect of bending/compression/ layers covering antenna
- effect of equipment in near-field

Antenna design constraints:

- Cost-effective, compact and low-profile for invisible integration
- wideband performance for high-datarate communication



→ a dedicated design strategy is needed!



Stochastic design framework for random variations

- effect of varying environmental conditions
- effect of fabrication tolerances
- effect of bending/compression/ layers covering antenna

Antenna topology with high antenna/platform isolation

- effect of equipment in near-field

Antenna design constraints:

- Cost-effective, compact and low-profile for invisible integration
- wideband performance for high-datarate communication



→ a dedicated design strategy

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- **Stochastic antenna design framework**
 - **Production uncertainties**
 - **Substrate compression**
 - **Substrate bending**
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Design strategies accounting for randomness

1. Overspecifying design requirements

→ enlarging bandwidth, applying stricter specs

☒ out-of-band interference

☒ cost

2. Quantifying random effects on antenna performance

→ applying Monte Carlo analysis

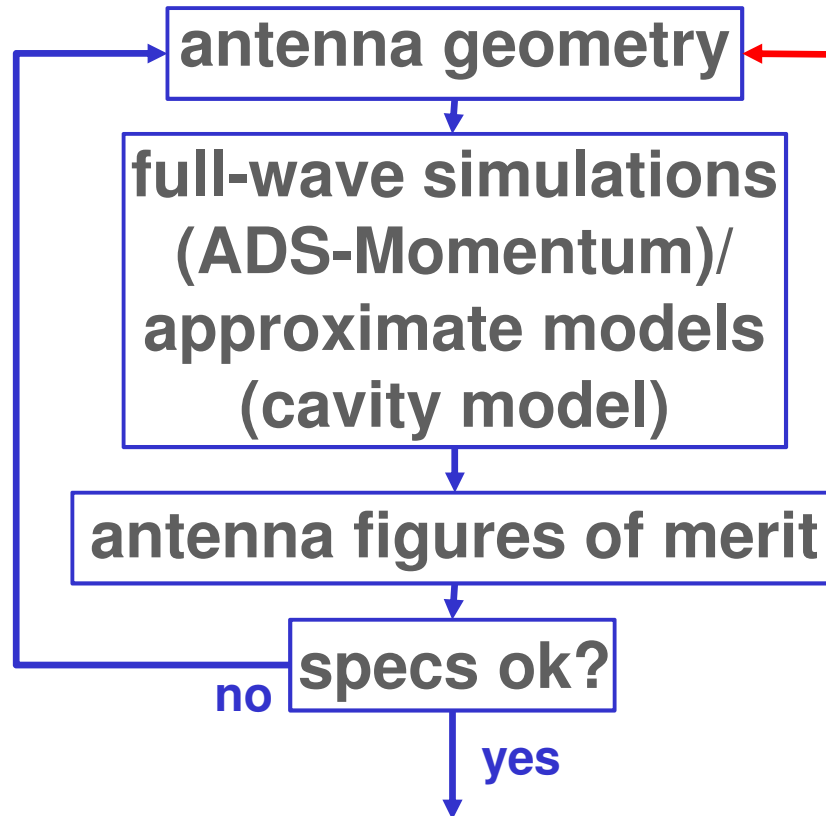
☑ very accurate

☒ time-consuming

→ a more effective stochastic formalism is needed!

Conventional design process

Stochastic design process



PDF antenna variations

antenna realizations

*limiting number of realizations
via
stochastic collocation method*

realizations figures of merit

PDF figures of merit

■ Polynomial chaos expansion $\mathbb{P}(X_1, X_2, \dots, X_N)$

- relates figures of merit Z to design parameters X_1, X_2, \dots, X_N

$$Z \approx \mathbb{P}(X_1, X_2, \dots, X_N) = \sum_{k_1=0}^{P_1} \sum_{k_2=0}^{P_2} \dots \sum_{k_N=0}^{P_N} y_{k_1 k_2 \dots k_N} \phi_{k_1}^{X_1}(X_1) \phi_{k_2}^{X_2}(X_2) \dots \phi_{k_N}^{X_N}(X_N)$$

- easy to find PDF figures of merit based on PDF design parameters
 - ♦ analytically based on polynomial
 - ♦ Monte Carlo applied to polynomial

■ Construction of $\mathbb{P}(X_1, X_2, \dots, X_N)$

$$Z \approx \mathbb{P}(X_1, X_2, \dots, X_N) = \sum_{k_1=0}^{P_1} \sum_{k_2=0}^{P_2} \dots \sum_{k_N=0}^{P_N} y_{k_1 k_2 \dots k_N} \phi_{k_1}^{X_1}(X_1) \phi_{k_2}^{X_2}(X_2) \dots \phi_{k_N}^{X_N}(X_N)$$

- statistically independent X_1, X_2, \dots, X_N
 - ◆ decorrelate random variables via Choleski decomposition
- for each random variable
 - ◆ basis function $\phi_{k_i}^{X_i}(X_i)$ orthonormal to input PDF $P_{X_i}(X_i)$

$$\langle \phi_i^X(X), \phi_j^X(X) \rangle = \int \phi_i^X(X) \phi_j^X(X) P_X(X) dX = \delta_{ij} = \begin{cases} 1 & \text{if } i = j \\ 0 & \text{if } i \neq j \end{cases}$$

■ Construction of $\mathbb{P}(X_1, X_2, \dots, X_N)$

$$Z \approx \mathbb{P}(X_1, X_2, \dots, X_N) = \sum_{k_1=0}^{P_1} \sum_{k_2=0}^{P_2} \dots \sum_{k_N=0}^{P_N} y_{k_1 k_2 \dots k_N} \phi_{k_1}^{X_1}(X_1) \phi_{k_2}^{X_2}(X_2) \dots \phi_{k_N}^{X_N}(X_N)$$

- exploiting orthonormality to calculate coefficients $y_{k_1 k_2 \dots k_N}$ via $V_1 \times V_2 \times \dots \times V_N$ -point Gaussian quadrature rule

$$y_{k_1 k_2 \dots k_N} = \int \dots \int_{\Gamma} \mathbb{F}(X_1, X_2, \dots, X_N) \phi_{k_1}^{X_1} \phi_{k_2}^{X_2} \dots \phi_{k_N}^{X_N} dP_{X_1, X_2, \dots, X_N}$$

$$\approx \sum_{l_1=0}^{V_1} \sum_{l_2=0}^{V_2} \dots \sum_{l_N=0}^{V_N} w_{l_1} w_{l_2} \dots w_{l_N} \phi_{k_1}^{X_1}(x_{l_1}) \phi_{k_2}^{X_2}(x_{l_2}) \dots \phi_{k_N}^{X_N}(x_{l_N}) \mathbb{F}(x_{l_1}', x_{l_2}', \dots, x_{l_N}')$$

x_{l_i} : zeros of $\phi_{k_i}^{X_i}$

weights w_{l_i} correspond to x_{l_i}

→ requires $V_1 \times V_2 \times \dots \times V_N$ realizations $\mathbb{F}(x_{l_1}', x_{l_2}', \dots, x_{l_N}')$

■ Alternative: Padé approximation

- Better for highly non-linear relationships:
rational function $f(X_1, X_2, \dots, X_N)$ instead of polynomial

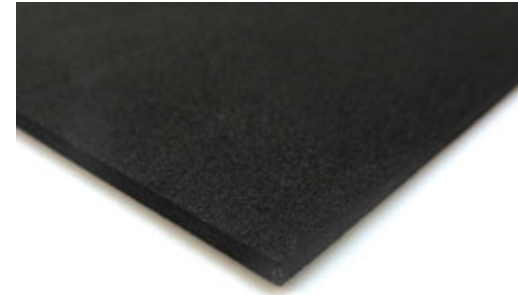
$$Z \approx f(X_1, X_2, \dots, X_N) = \frac{P_M(X_1, X_2, \dots, X_N)}{Q_L(X_1, X_2, \dots, X_N)}$$

with polynomials $P_M(X_1, X_2, \dots, X_N) = \sum_{j=0}^{c(M)} p_j \Phi_j$

$$Q_L(X_1, X_2, \dots, X_N) = \sum_{j=0}^{c(L)} q_j \Phi_j, \quad Q_L \neq 0 \text{ in } \Gamma$$

1. Production uncertainties

- non-uniformity in textile substrates
- variations in patch geometry
- feed misplacement

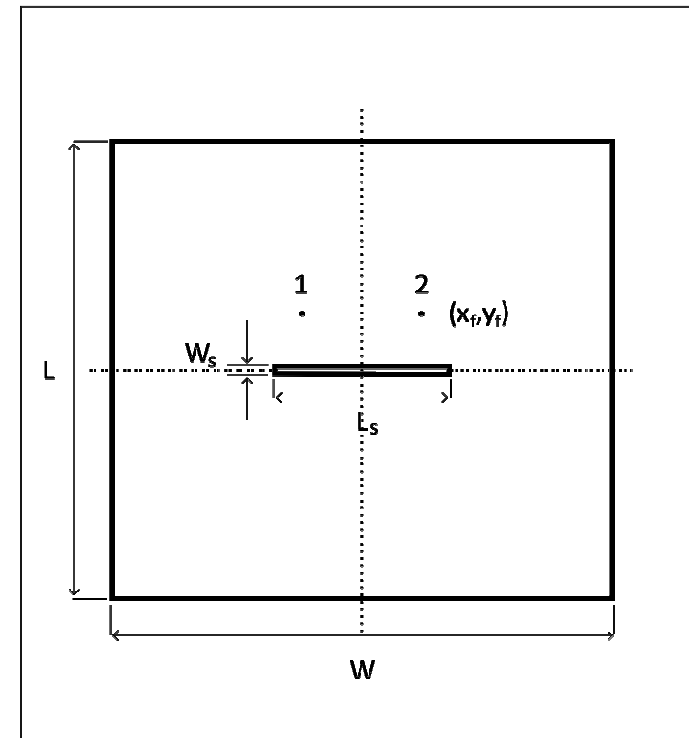


■ probe-fed 2.45GHz ISM-band patch antenna

- nominal input impedance

$$Z_{in} = 50 \Omega \text{ at } 2.45\text{GHz}$$

W	44.46 mm
L	45.32 mm
(x_f, y_f)	$(\pm 5.7, 5.7)$ mm
W_s	1 mm
L_s	14.88 mm



protective foam substrate ($\epsilon_r=1.53$, $h=3.94$ mm)



- variations in patch width W : largest influence on Z_{in}

- measurements on 100 patches, manually cut

- ◆ mean value

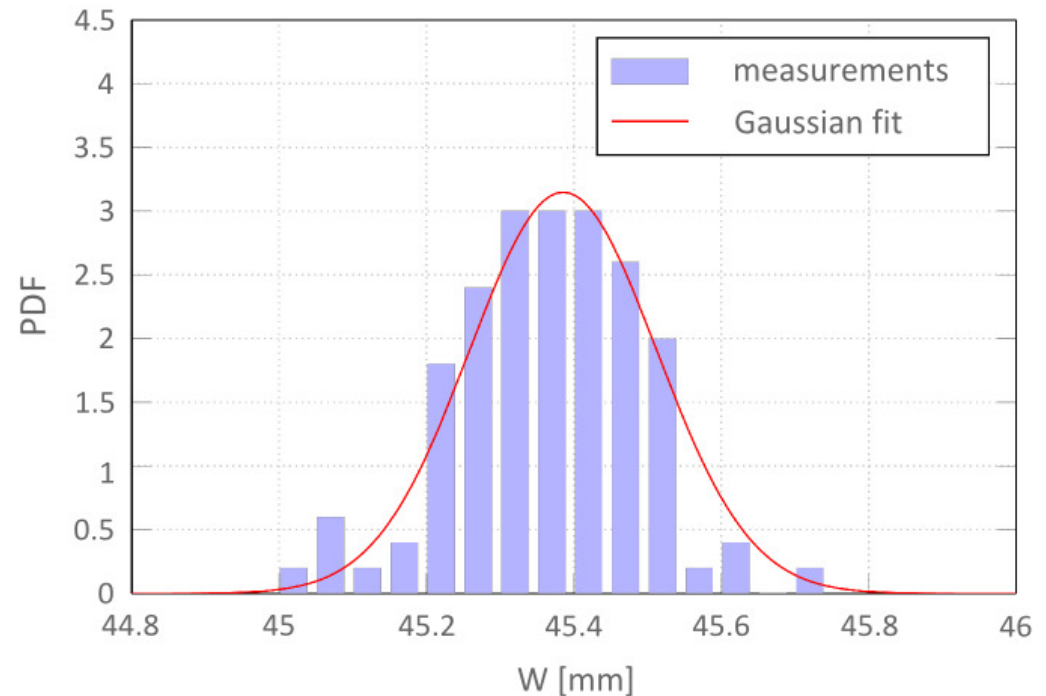
$$\bar{W} = 45.385 \text{ mm}$$

- ◆ standard deviation

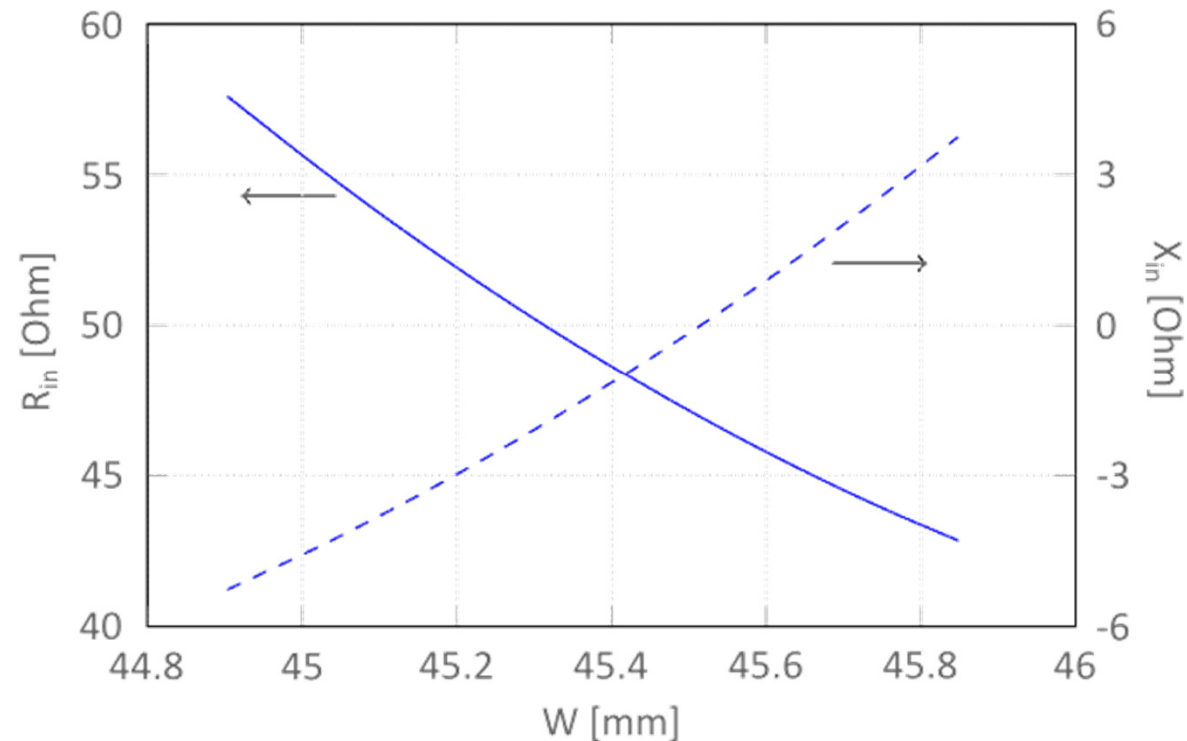
$$\sigma = 0.127$$

- ◆ variation interval

$$[44.9 - 45.9] \text{ mm}$$



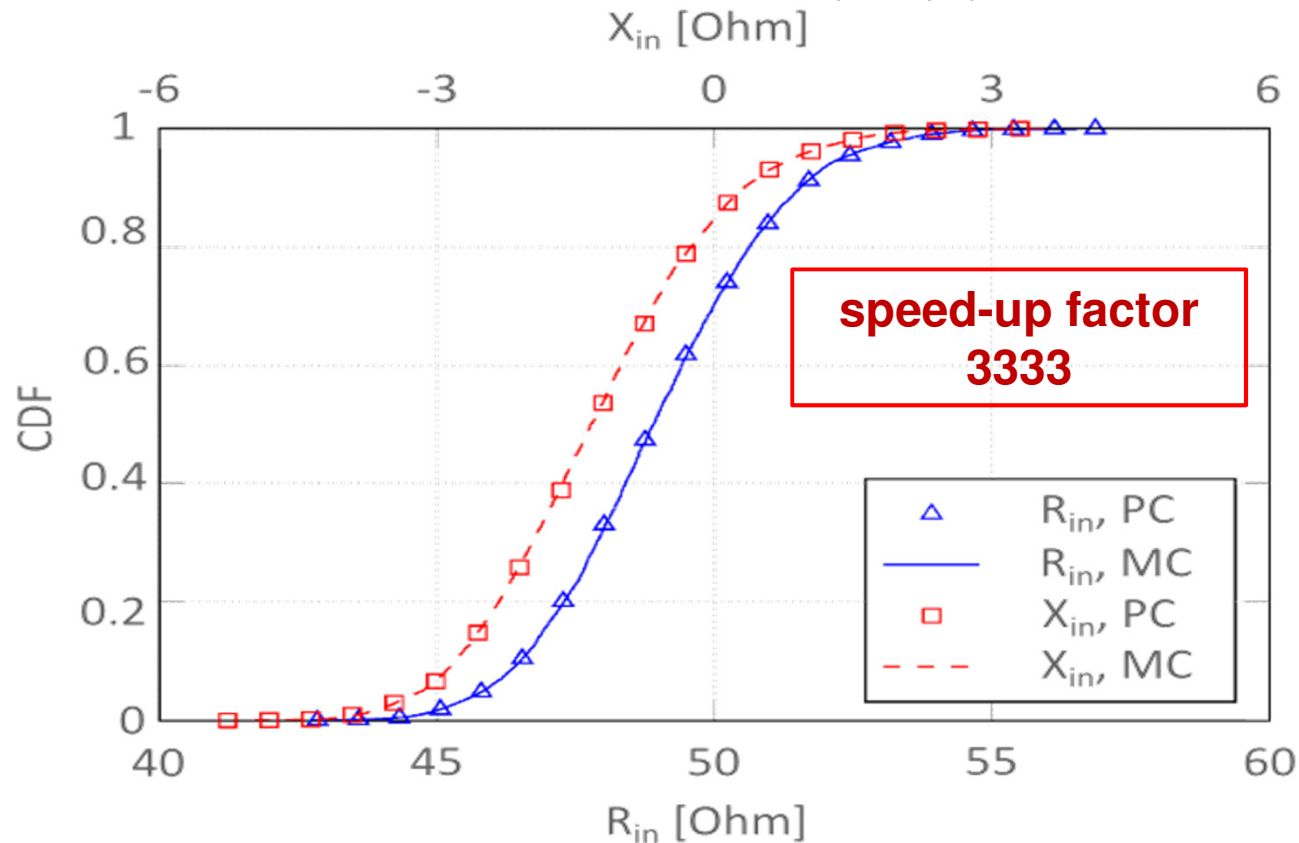
- relates patch width W to Z_{in}
 - convergence for polynomial order $P = 2$
 - $V = 3$ quadrature points



simulation time: **18 s** (Intel Core i7-2600, 3.40 GHz, 16 GB RAM)

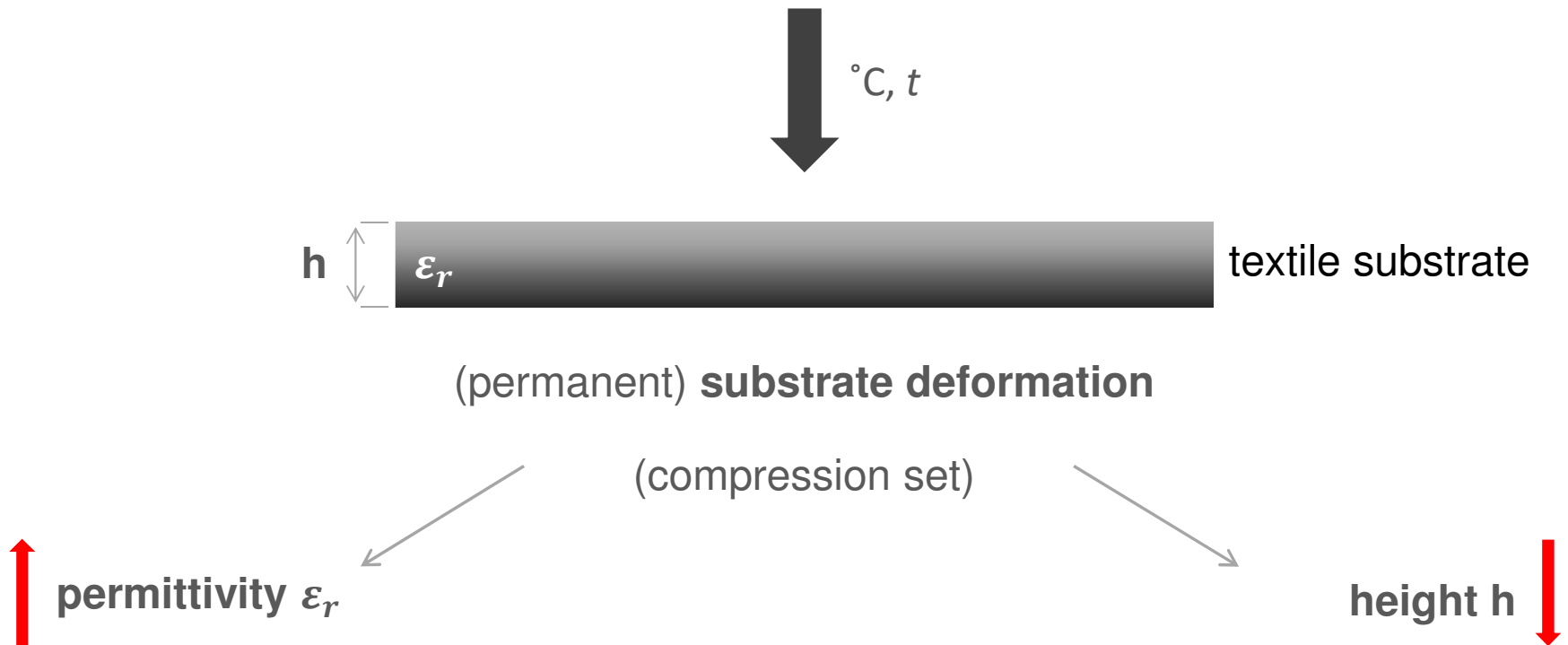
■ Output PDF of Z_{in} generated with 10000 realizations

- Monte-Carlo based on polynomial expansion (PC) (CPU-time 18s) versus based on full-wave simulations (MC) (CPU-time 16h 40min)



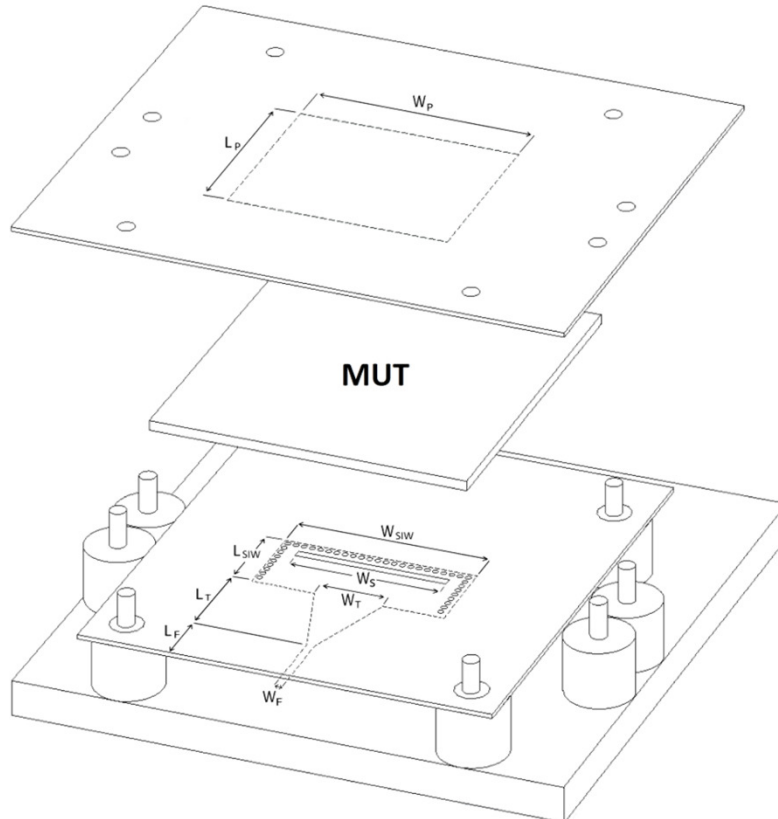
2. Substrate compression

- 2D stochastic model for correlated height and permittivity



2. Substrate compression: permittivity measurement

- 2D model for correlated height and permittivity: input PDF



Resonance perturbation method based on aperture-coupled patch antenna:

- minimized interference and back radiation
- highest field strength confined in the material under test (MUT)
- fast replacement of MUT

ϵ_r extracted by comparing measurements to CST Microwave Studio simulations

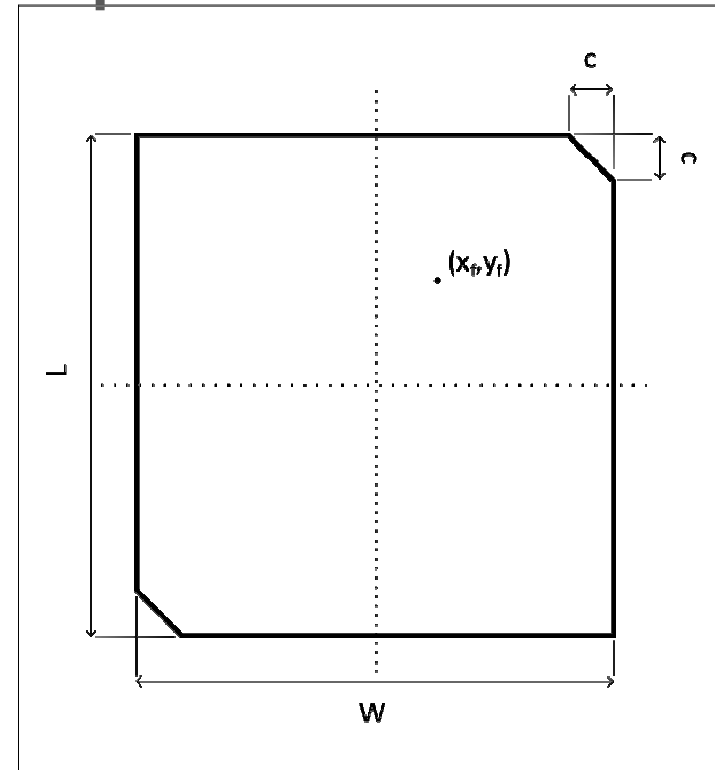
■ probe-fed microstrip GPS L1-band patch antenna

- requirements L1-band (1.57GHz):

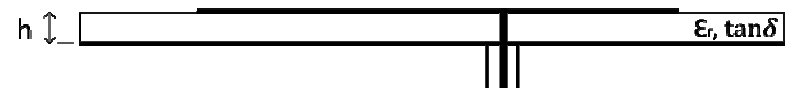
$$|S_{11}| \leq 0.316 \text{ } (-10\text{dB})$$

$$AR \leq 1.41 \text{ } (3\text{dB}) \text{ (RHCP)}$$

L	73.5 mm
W	69.5 mm
(x_f, y_f)	(8.5, 14.5) mm
c	5 mm

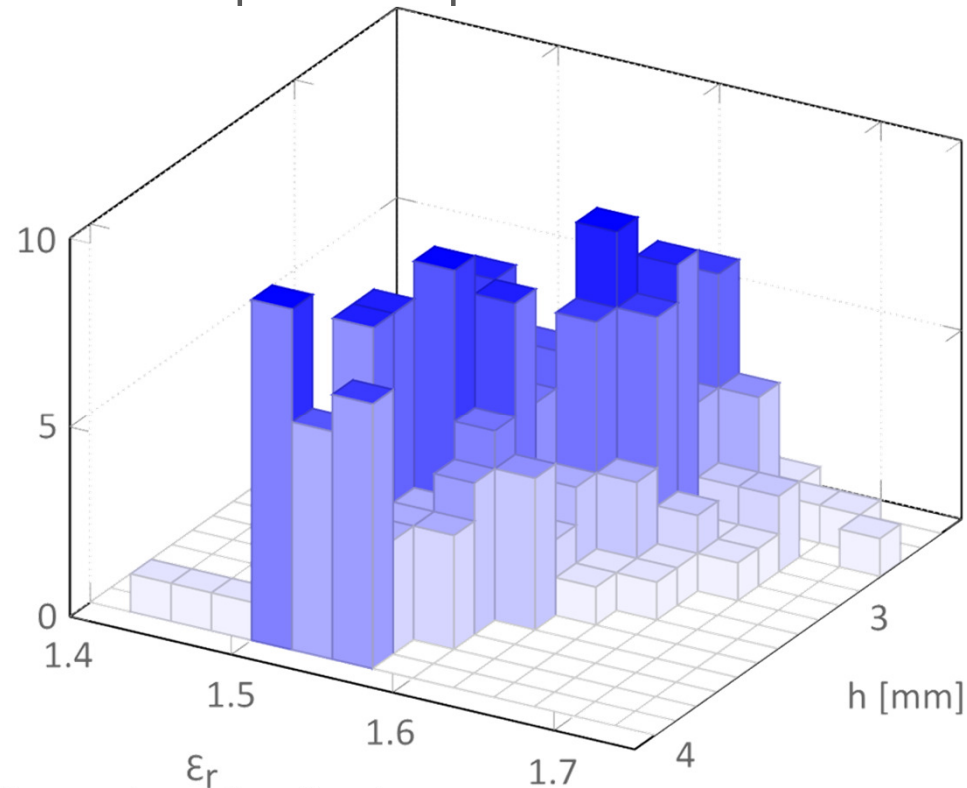


protective foam substrate ($\epsilon_r=1.53, h=3.94$ mm)



■ Input PDF

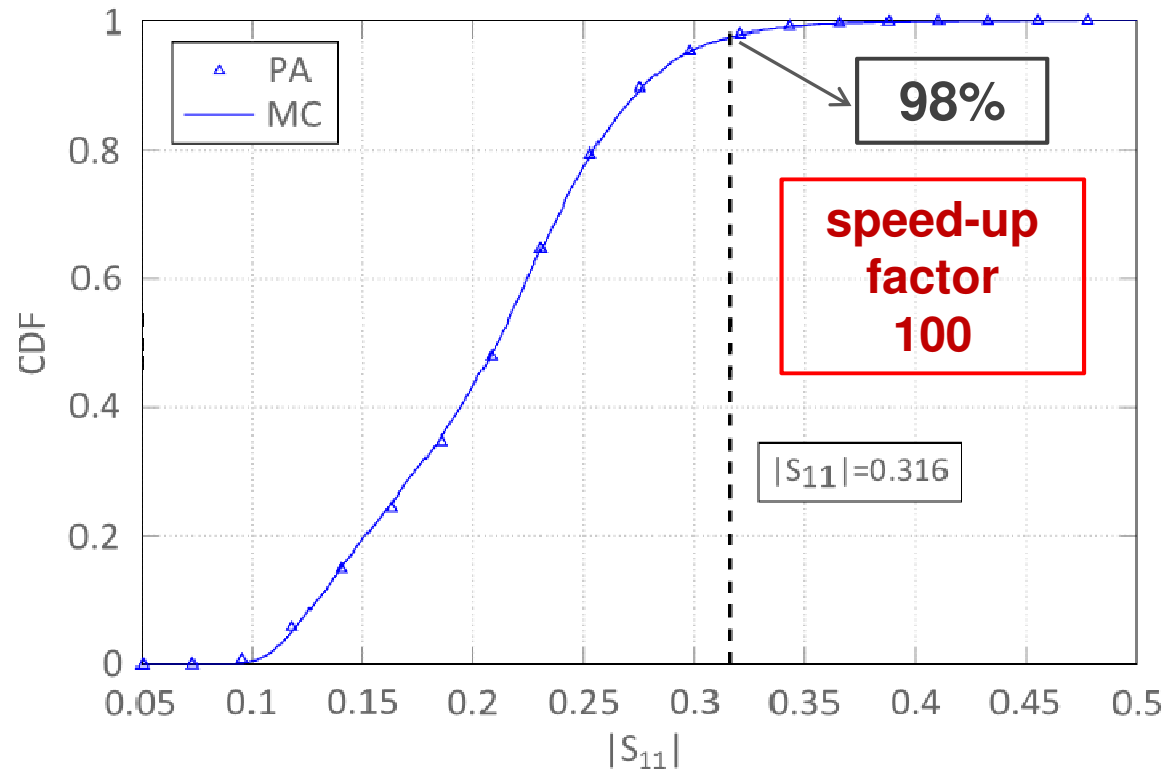
- 25 foam substrate samples compressed and measured 10 times



- fit to bivariate Gaussian distribution
 - ◆ $\mu_h = 3.36 \text{ mm}$, $\mu_{\epsilon_r} = 1.58$, $\sigma_h = 0.337$, $\sigma_{\epsilon_r} = 0.049$, correlation $\rho = -0.68$

■ Output PDF constructed by Padé-chaos expansion

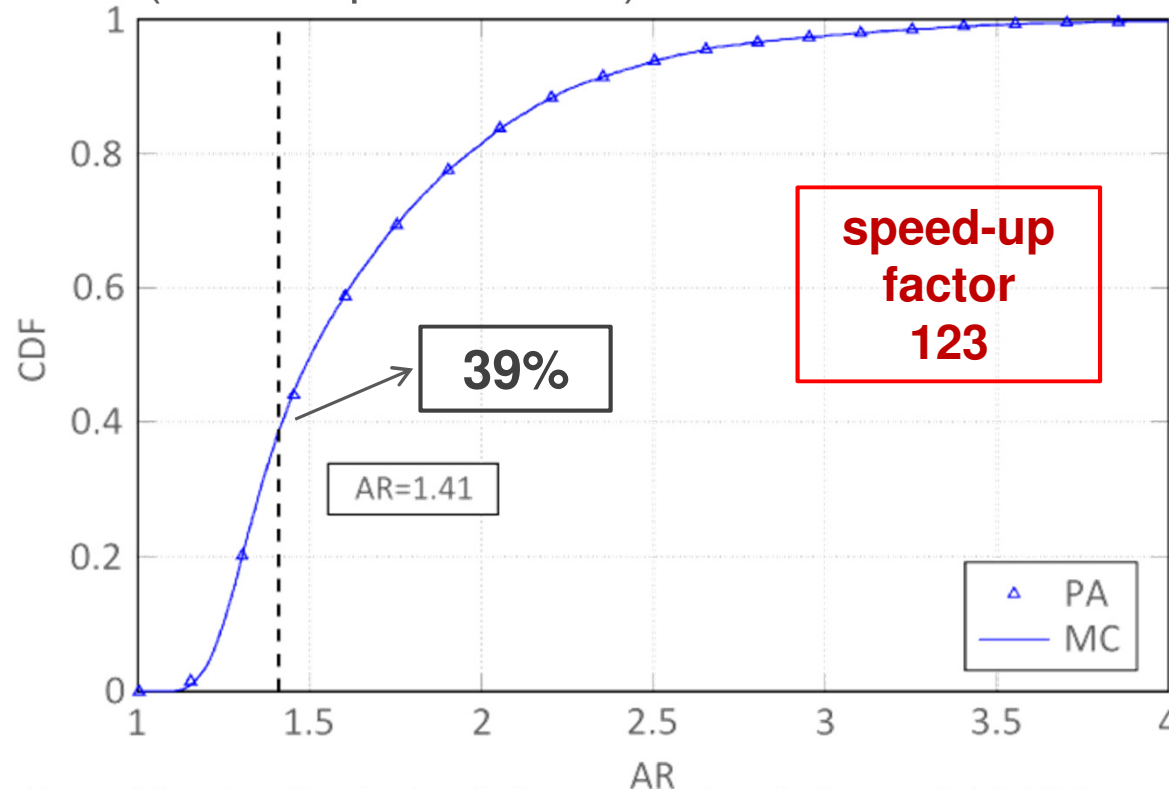
- antenna return loss at 1.57GHz



- validation: Monte-Carlo by full-wave simulation of 10000 realizations
 - ◆ simulation time: 44 h 26 m

■ Output PDF constructed by Padé-chaos expansion

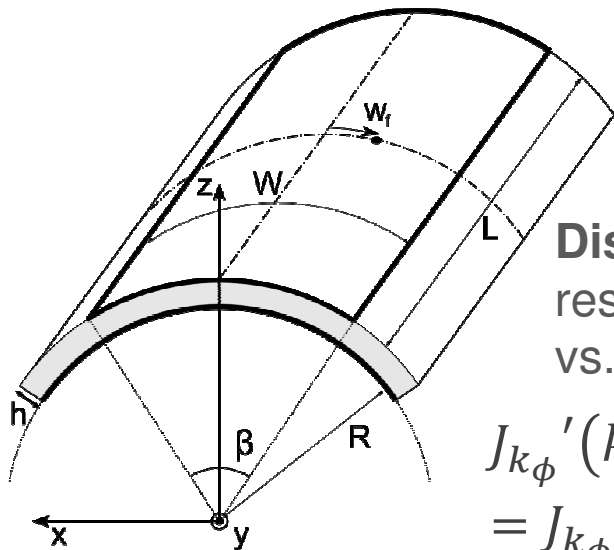
- axial ratio (circular polarization) at 1.57GHz



- validation: Monte-Carlo by full-wave simulation of 10000 realizations
 - ◆ simulation time: 44 h 26 m

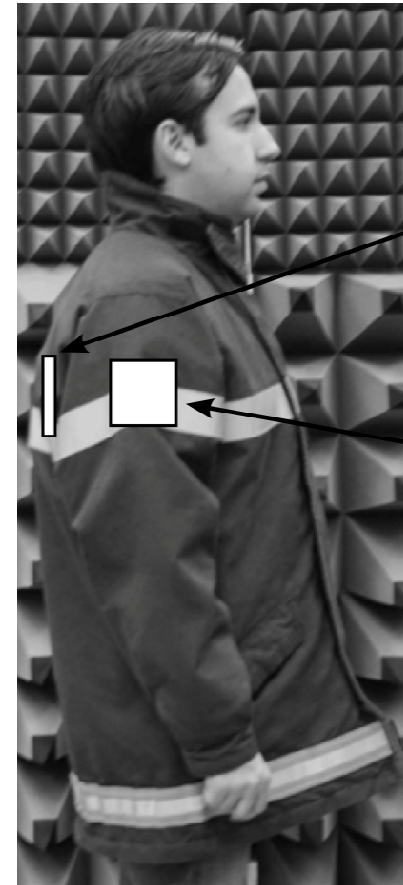
3. Substrate bending

- variations in antenna's curvature radius
 - ♦ movements and activity of person wearing antenna
 - ♦ variations in body morphology over many different users
 - ♦ modeled by cavity model



Dispersion relation:
resonance frequency f_r
vs. bending radius R

$$J_{k_\phi}'(k_p R) Y_{k_\phi}'(k_p (R + h)) = J_{k_\phi}'(k_p (R + h)) Y_{k_\phi}'(k_p R)$$



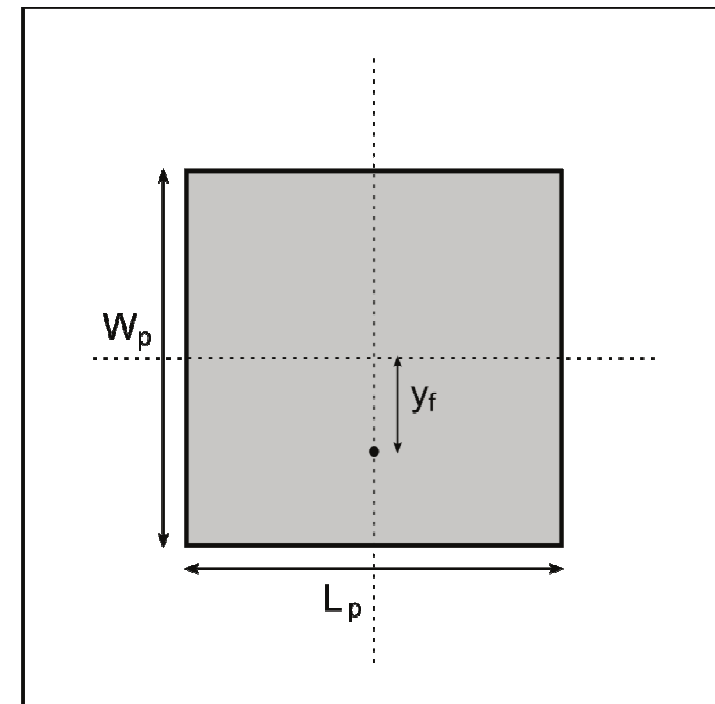
antenna planar
on-body position

antenna bent
on-body position

■ probe-fed microstrip GPS L1-band patch antenna

- Nominal resonance frequency $f_r = 1.57\text{GHz}$

W_p	69.25 mm
L_p	81.2 mm
y_f	16 mm



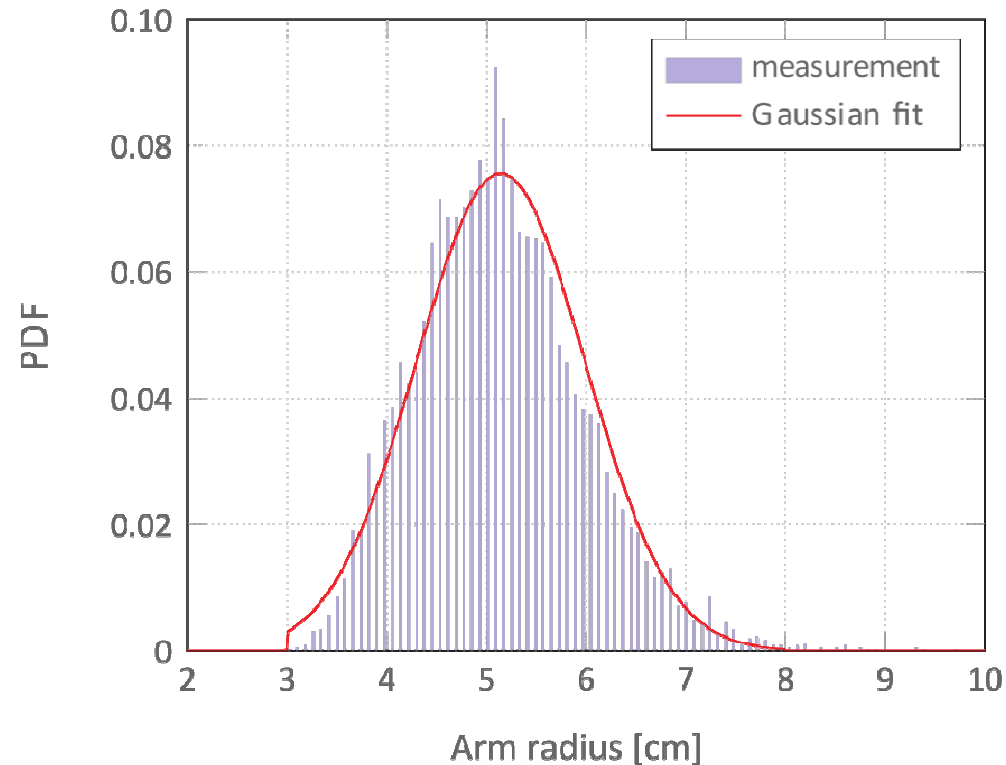
copper foil antenna patch ($R_s = 0.18 \Omega/sq$)

aramid fabric ($\epsilon_r=1.75$, $h=2$ mm)



■ Input PDF arm radius R of human population

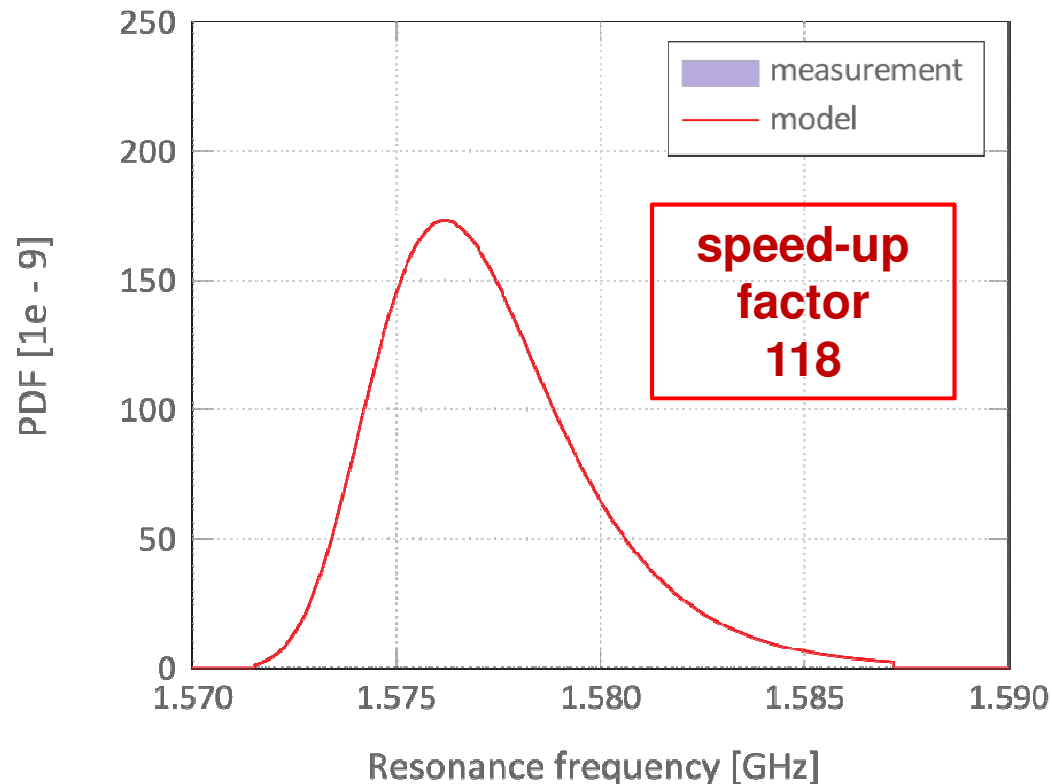
- from NHANES database



- fit to truncated Gaussian distribution
 - ◆ $\bar{R} = 5.14 \text{ cm}$, $\sigma = 0.85$, variation interval [3–8] cm

■ Output PDF by polynomial chaos expansion

- antenna resonance frequency

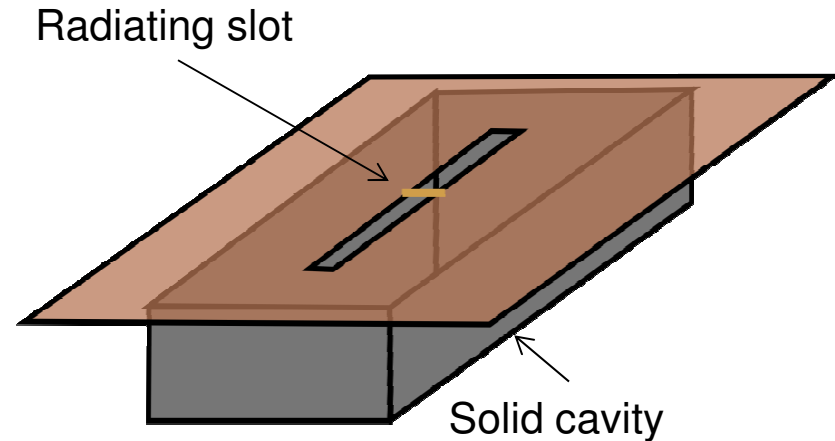


- validation: Monte-Carlo by full-wave simulation of 7056 realizations
 - ◆ simulation time: 59s

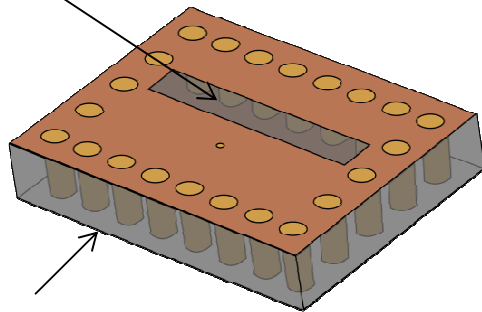
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 - **Substrate-Integrated-Waveguide (SIW)
Cavity-Backed Slot (CBS) topology**
 - **Three-element antenna array for integration into furniture**
 - **Half-mode SIW CBS antenna on cork substrate**
- **Conclusions**

Cavity-backed slot antenna topology

- High radiation efficiency
- High front-to-back ratio
- High isolation from its environment



Radiating slot



SIW cavity

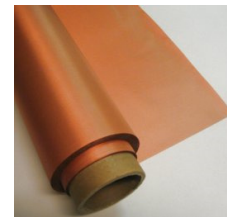
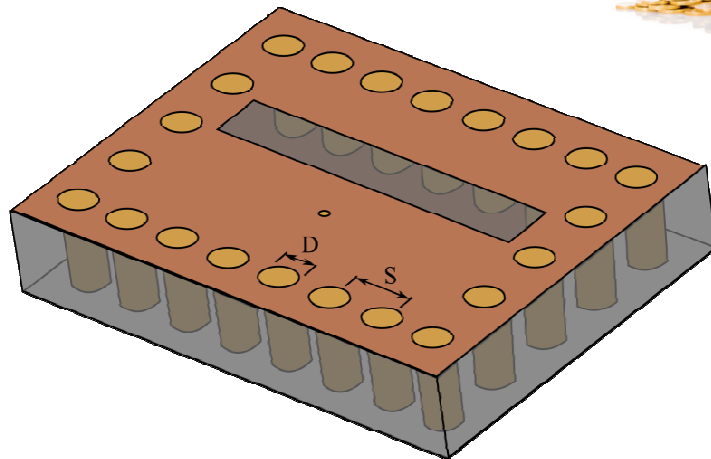
Substrate Integrated Waveguide (SIW) technology

- low-profile
- cost-effective
- simple implementation
- easy integration with planar circuitry

Antenna materials

Application-specific antenna substrate

- Stable, high-performance
- Green, recyclable design
- Reuse object's material if possible
- Significant cost and area reduction
- Reliable material characterization necessary



Copper-plated tafetta

- Conductive layers
- $R_s = 0.18 \Omega/\text{sq}$



Brass tubular eyelets

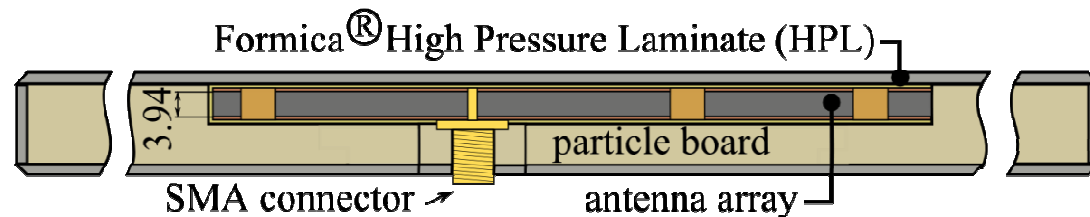
- Effective electric walls
- Closely spaced to minimize radiation loss ($S/D < 2.5$)

Goal

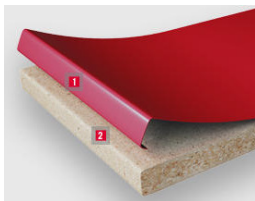
- Ultra-wideband three-element SIW CBS antenna array
- Invisible integration into furniture (desks, worktop kitchen, ...)



Integration procedure

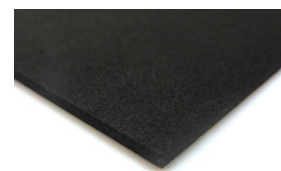


Integration platform



- Common material for furniture
- 38 mm-thick particle board
- 1 mm-thick High Pressure Laminate (HPL)
- $\epsilon_r = 2.7 - 3.07$ [1]
- $\tan\delta = 0.07 - 0.09$ [1]

Antenna substrate



- 3.94 mm-thick closed-cell expanded-rubber
- Low losses
- Low moisture regain
- $\epsilon_r = 1.495$ @ 5.50 GHz [2]
- $\tan\delta = 0.016$ @ 5.50 GHz [2]

[1] G. I. Torgovnikov, Dielectric Properties of Wood and Wood-based Materials, Springer-Verlag, Berlin, 1993

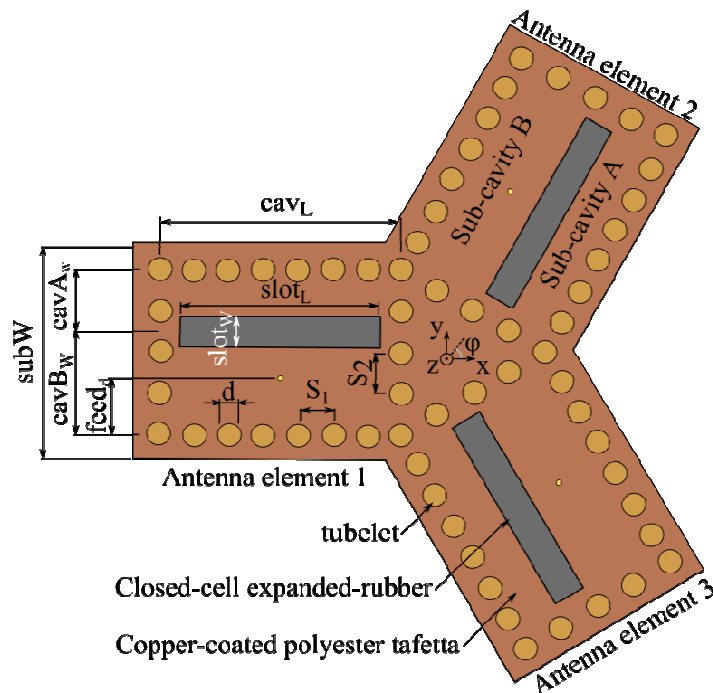
[2] F. Declercq, H. Rogier, and C. Hertleer, "Permittivity and Loss Tangent Characterization for Garment Antennas Based on a New Matrix-Pencil Two-Line Method," IEEE Trans. Antennas Propagat., vol. 56, no. 8, pp. 2548–2554, Aug 2008.

Design requirements

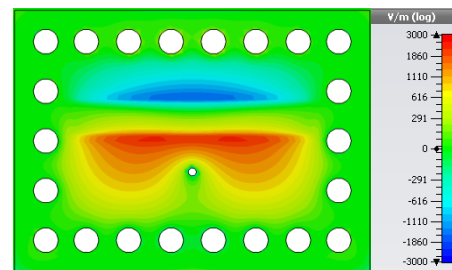
- 5 GHz Wi-Fi band [5.15-5.85] GHz , with 250 MHz margins
 - Return loss > 10 dB, [4.90-6.10] GHz
 - Isolation > 25 dB, [4.90-6.10] GHz

Single element design

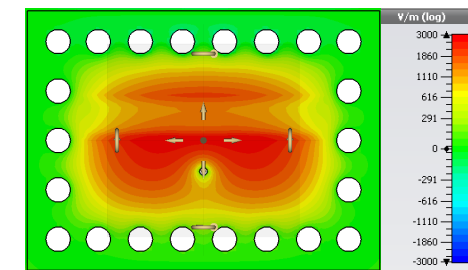
- Coaxial feed
- Linearly-polarized
- Multi-moded bandwidth enhancement technique [3]



Field distribution



lowest resonance frequency

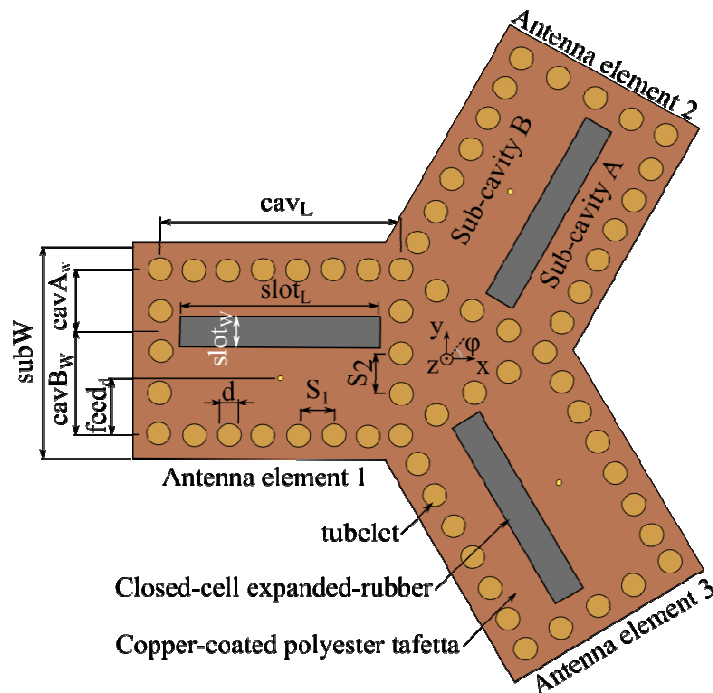


highest resonance frequency

[3] G. Q. Luo, Z. F. Hu, W. J. Li, X. H. Zhang, L. L. Sun, and J. F. Zheng, "Bandwidth-Enhanced Low-Profile Cavity-Backed Slot Antenna by Using Hybrid SIW Cavity Modes," IEEE Trans. Antennas Propagat., vol. 60, no. 4, pp. 1698–1704, Apr. 2012.

Design requirements

- 5 GHz Wi-Fi band [5.15-5.85] GHz , with 250 MHz margins
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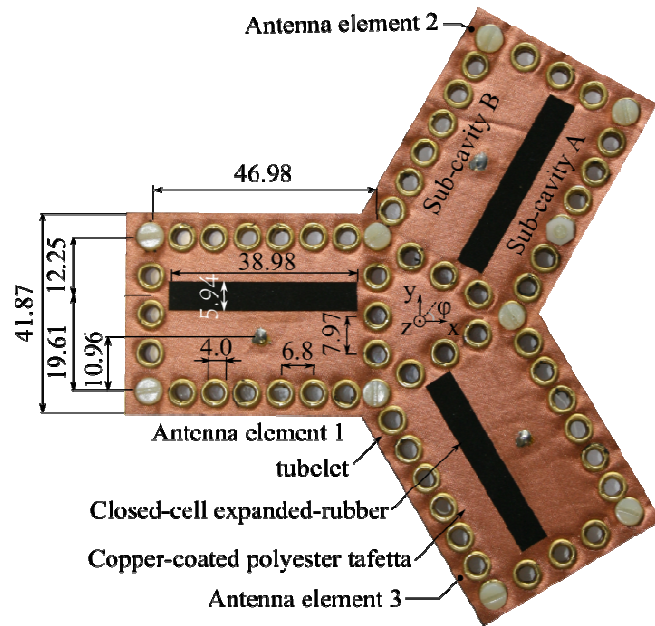


Array design

- Three identical antenna elements
- Threefold rotational symmetry
- Polarization diversity
- Spatial diversity

➔ High MIMO/diversity gain

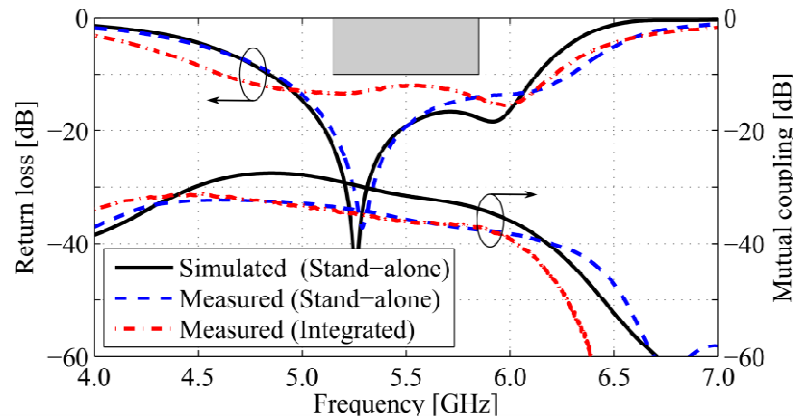
Prototype



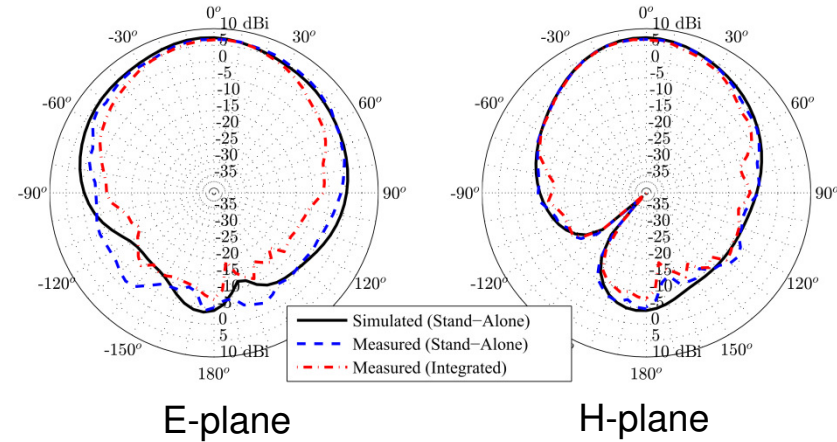
Invisible integration in conference table



S-parameters



Radiation (far-field) performance

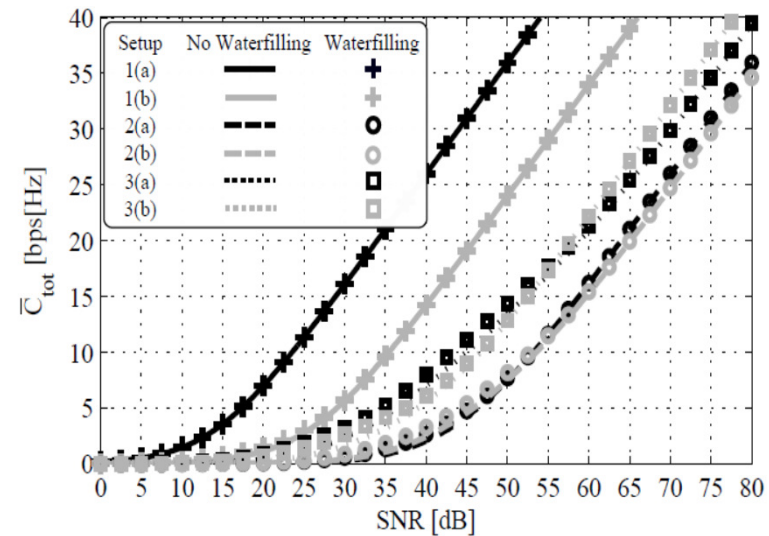
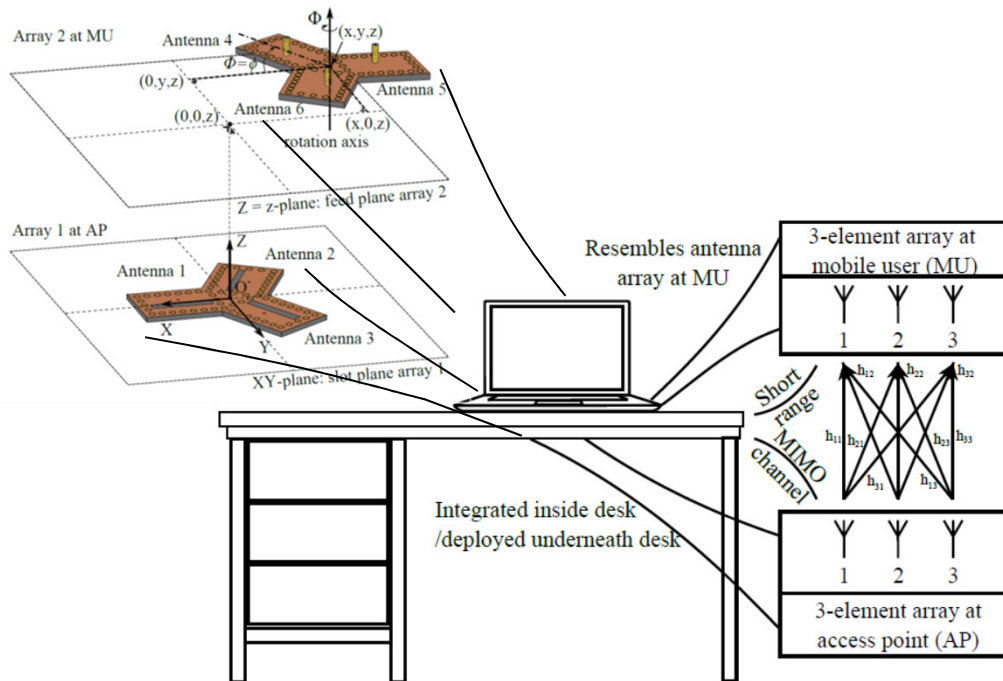


Summary

	Simulated (Stand-Alone)	Measured (Stand-Alone)	Measured (Integrated)
Impedance bandwidth [MHz]	1296	1433	1652
η_{rad} at 5.5 GHz [%]	91	84	78
Gain _{max} at 5.5 GHz [dBi]	6.45	5.9	5.8
FTBR at 5.5 GHz [dB]	13.2	10.9	13.2

■ ultra-high datarate short-distance channel

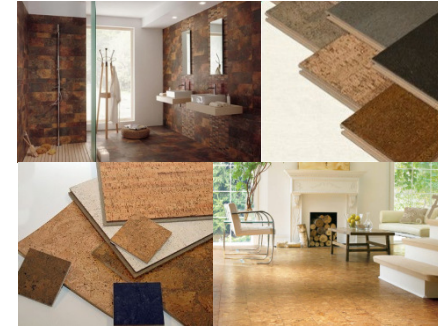
- 3x3 spatial multiplexing 802.11ac link
- wideband antenna array in desk and laptop



- ◆ up to 4.2 Gbps at 40dB SNR in 160MHz bandwidth
- ◆ up to 18.2 Gbps at 40dB SNR in 700MHz bandwidth

Goal

- Miniaturized, wideband single-element SIW CBS antenna
- Invisible integration into cork floor and wall tiles



Antenna substrate



- 3 mm-thick cork substrate by Amorim Cork Composites S.A.
- Cork granules, bound by polyurethane
- Invisible integration: similar substrate as superstrate
- Low losses, low moisture regain
- Characterized at 5.50 GHz using resonator technique [5]

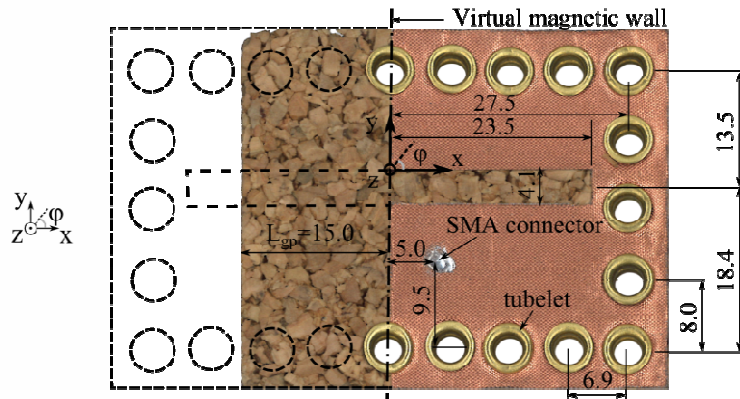
Antenna design

- Based on average parameters $(\epsilon_r, \tan_d) = (1.22, 0.0363)$
- Retain 300 MHz impedance bandwidth margins
- Allow for slight variation in cork material properties

➔ Return loss > 10 dB, [4.85-6.15] GHz

Resonator	(ϵ_r, \tan_d)
1	(1.20, 0.0306)
2	(1.17, 0.0244)
3	(1.25, 0.0481)
4	(1.26, 0.0421)

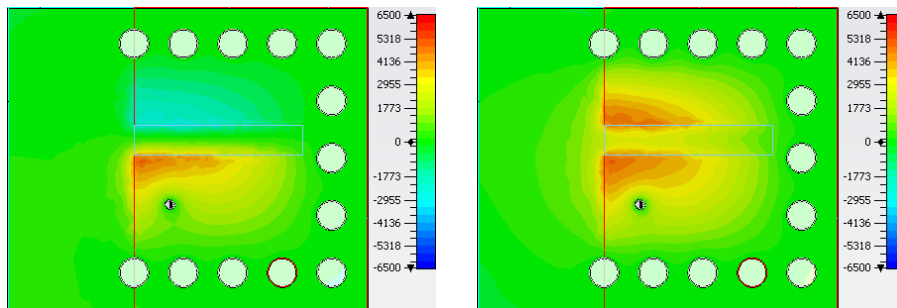
[5] O. Caytan, S. Lemey, S. Agneessens, D. Vande Ginste, P. Demeester, C. Loss, R. Salvado, and H. Rogier, "Half-mode substrate-integrated-waveguide cavity-backed slot antenna on cork substrate," *Antennas Wirel. Propag. Lett.*, vol. PP, no. 99, 2015.



Half-mode SIW CBS Antenna

- Return loss > 10 dB, [4.85-6.15] GHz
- Multi-moded bandwidth enhancement technique
- Half-mode SIW miniaturization technique

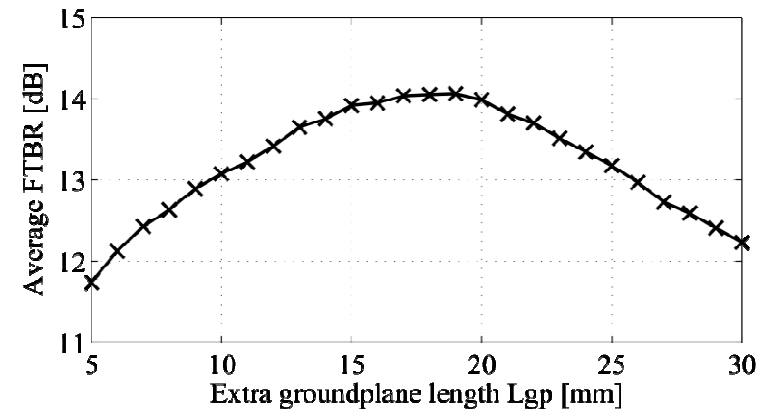
Field distribution



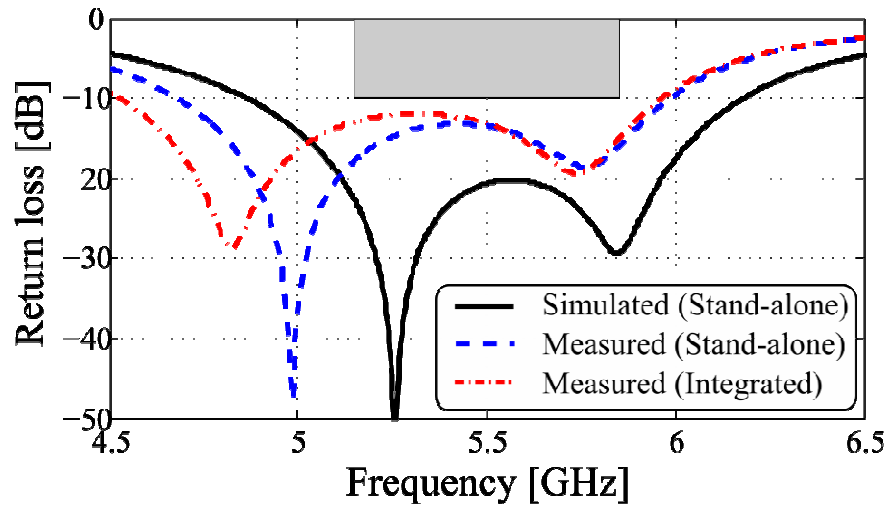
lowest resonance frequency

highest resonance frequency

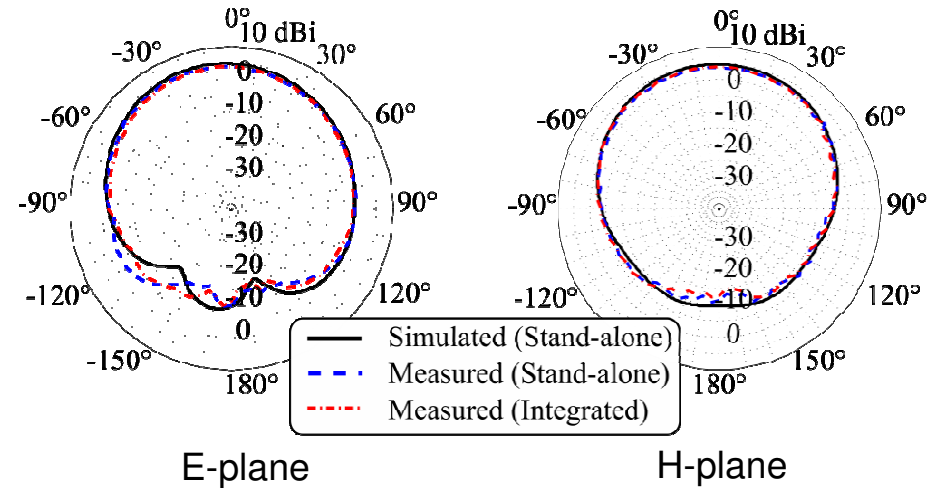
Optimal Front-to-back ratio (FTBR)



Return loss



Radiation performance at 5.5 GHz



Summary

	Simulated (Stand-Alone)	Measured (Stand-Alone)	Measured (Integrated)
Impedance bandwidth [MHz]	1317	1312	1436
η_{rad} at 5.5 GHz [%]	83	85	80
Gain _{max} at 5.5 GHz [dBi]	5.0	4.3	4.2
FTBR at 5.5 GHz [dB]	14.5	15.0	16.8

- **The Internet-of-Things**
 - opportunities and design challenges
- **Stochastic antenna design framework**
 - Production uncertainties
 - Substrate compression
 - Substrate bending
- **Some novel IoT antenna designs**
 - Substrate-Integrated-Waveguide (SIW)
Cavity-Backed Slot (CBS) topology
 - Three-element antenna array for integration into furniture
 - Half-mode SIW CBS antenna on cork substrate
- **Conclusions**

■ Accounting for random variations in IoT antennas

- in antenna geometry
- In deployment conditions

→ Stochastic antenna design framework

- quantify statistics of antenna's figures of merit

■ Better isolation between antenna and IoT platform

→ Dedicated antenna topology

- Cavity-backed slot antenna in SIW technology

■ Design examples

- three-element UWB antenna array for integration into furniture
- miniature HMSIW antenna for integration in cork floors and walls

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